

6.3 A coil consists of 100 turns of wire wrapped around a square frame of sides 0.25 m. The coil is centered at the origin with each of its sides parallel to the x - or y -axis. Find the induced emf across the open-circuited ends of the coil if the magnetic field is given by

(a) $\mathbf{B} = \hat{\mathbf{z}}20e^{-3t}$ (T)

(b) $\mathbf{B} = \hat{\mathbf{z}}20\cos x \cos 10^3t$ (T)

(c) $\mathbf{B} = \hat{\mathbf{z}}20\cos x \sin 2y \cos 10^3t$ (T)

Solution: Since the coil is not moving or changing shape, $V_{\text{emf}}^m = 0$ V and $V_{\text{emf}} = V_{\text{emf}}^{\text{tr}}$. From Eq. (6.6),

$$V_{\text{emf}} = -N \frac{d}{dt} \int_S \vec{\mathbf{B}} \cdot d\vec{\mathbf{s}} = -N \frac{d}{dt} \int_{-0.125}^{0.125} \int_{-0.125}^{0.125} \vec{\mathbf{B}} \cdot (\hat{\mathbf{z}} dx dy),$$

where $N = 100$ and the surface normal was chosen to be in the $+\hat{\mathbf{z}}$ direction.

(a) For $\vec{\mathbf{B}} = \hat{\mathbf{z}}20e^{-3t}$ (T),

$$V_{\text{emf}} = -100 \frac{d}{dt} \left(20 \exp -3t (0.25)^2 \right) = 375e^{-3t} \quad (\text{V}).$$

(b) For $\vec{\mathbf{B}} = \hat{\mathbf{z}}20\cos x \cos 10^3t$ (T),

$$V_{\text{emf}} = -100 \frac{d}{dt} \left(20 \cos 10^3t \int_{x=-0.125}^{0.125} \int_{y=-0.125}^{0.125} \cos x dx dy \right) = 124.6 \sin 10^3t \quad (\text{kV}).$$

(c) For $\vec{\mathbf{B}} = \hat{\mathbf{z}}20\cos x \sin 2y \cos 10^3t$ (T),

$$V_{\text{emf}} = -100 \frac{d}{dt} \left(20 \cos 10^3t \int_{x=-0.125}^{0.125} \int_{y=-0.125}^{0.125} \cos x \sin 2y dx dy \right) = 0.$$

6.5 A circular-loop TV antenna with 0.02-m^2 area is in the presence of a uniform-amplitude 300-MHz signal. When oriented for maximum response, the loop develops an emf with a peak value of 30 (mV). What is the peak magnitude of \mathbf{B} of the incident wave?

Solution: TV loop antennas have one turn. At maximum orientation, Eq. (6.5) evaluates to $\Phi = \int \vec{\mathbf{B}} \cdot d\vec{\mathbf{s}} = \pm BA$ for a loop of area A and a uniform magnetic field with magnitude $B = |\mathbf{B}|$. Since we know the frequency of the field is $f = 300$ MHz, we can express B as $B = B_0 \cos(\omega t + \alpha_0)$ with $\omega = 2\pi \times 300 \times 10^6$ rad/s and α_0 an arbitrary reference phase. From Eq. (6.6),

$$V_{\text{emf}} = -N \frac{d\Phi}{dt} = -A \frac{d}{dt} [B_0 \cos(\omega t + \alpha_0)] = AB_0 \omega \sin(\omega t + \alpha_0).$$

V_{emf} is maximum when $\sin(\omega t + \alpha_0) = 1$. Hence,

$$30 \times 10^{-3} = AB_0 \omega = 0.02 \times B_0 \times 6\pi \times 10^8,$$

which yields $B_0 = 0.8$ (nT).

6.11 The loop shown in P6.11 moves away from a wire carrying a current $I_1 = 10$ A at a constant velocity $\mathbf{u} = \hat{\mathbf{y}}7.5$ (m/s). If $R = 10 \Omega$ and the direction of I_2 is as defined in the figure, find I_2 as a function of y_0 , the distance between the wire and the loop. Ignore the internal resistance of the loop.

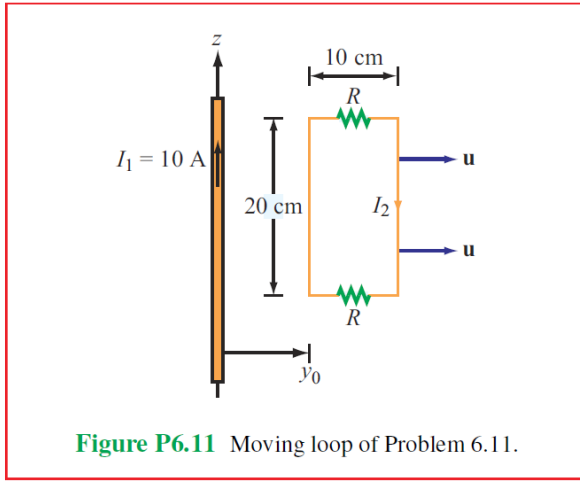


Figure P6.11 Moving loop of Problem 6.11.

Solution: Assume that the wire carrying current I_1 is in the same plane as the loop. The two identical resistors are in series, so $I_2 = V_{\text{emf}}/2R$, where the induced voltage is due to motion of the loop and is given by Eq. (6.26):

$$V_{\text{emf}} = V_{\text{emf}}^{\text{m}} = \oint_C (\vec{u} \times \vec{B}) \cdot d\vec{l}.$$

The magnetic field \vec{B} is created by the wire carrying I_1 . Choosing $\hat{\mathbf{z}}$ to coincide with the direction of I_1 , Eq. (5.30) gives the external magnetic field of a long wire to be

$$\vec{B} = \hat{\phi} \frac{\mu_0 I_1}{2\pi r}.$$

For positive values of y_0 in the y - z plane, $\hat{\mathbf{y}} = \hat{\mathbf{r}}$, so

$$\vec{u} \times \vec{B} = \hat{\mathbf{y}}|\vec{u}| \times \vec{B} = \hat{\mathbf{r}}|\vec{u}| \times \hat{\phi} \frac{\mu_0 I_1}{2\pi r} = \hat{\mathbf{z}} \frac{\mu_0 I_1 u}{2\pi r}.$$

Integrating around the four sides of the loop with $d\vec{l} = \hat{\mathbf{z}} dz$ and the limits of integration chosen in accordance with the assumed direction of I_2 , and recognizing that only the two sides without the resistors contribute to $V_{\text{emf}}^{\text{m}}$, we have

$$V_{\text{emf}}^{\text{m}} = \int_0^{0.2} \left(\hat{\mathbf{z}} \frac{\mu_0 I_1 u}{2\pi r} \right) \Big|_{r=y_0} \cdot (\hat{\mathbf{z}} dz) + \int_{0.2}^0 \left(\hat{\mathbf{z}} \frac{\mu_0 I_1 u}{2\pi r} \right) \Big|_{r=y_0+0.1} \cdot (\hat{\mathbf{z}} dz)$$

$$= \frac{4\pi \times 10^{-7} \times 10 \times 7.5 \times 0.2}{2\pi} \left(\frac{1}{y_0} - \frac{1}{y_0 + 0.1} \right)$$

$$= 3 \times 10^{-6} \left(\frac{1}{y_0} - \frac{1}{y_0 + 0.1} \right) \quad (\text{V}),$$

and therefore

$$I_2 = \frac{V_{\text{emf}}^{\text{m}}}{2R} = 150 \left(\frac{1}{y_0} - \frac{1}{y_0 + 0.1} \right) \quad (\text{nA}).$$

6.17 In wet soil, characterized by $\sigma = 10^{-2}$ (S/m), $\mu_r = 1$, and $\epsilon_r = 36$, at what frequency is the conduction current density equal in magnitude to the displacement current density?

Solution: For sinusoidal wave variation, the phasor electric field is

$$\begin{aligned} E &= E_0 e^{j\omega t} \\ J_c &= \sigma E = \sigma E_0 e^{j\omega t} \\ J_d &= \frac{\partial D}{\partial t} = \epsilon \frac{\partial E}{\partial t} = j\omega \epsilon E_0 e^{j\omega t} \\ \left| \frac{J_c}{J_d} \right| &= 1 = \frac{\sigma}{\omega \epsilon} = \frac{\sigma}{2\pi \epsilon f} \end{aligned}$$

or

$$f = \frac{\sigma}{2\pi \epsilon} = \frac{10^{-2}}{2\pi \times 36 \times 8.85 \times 10^{-12}} = 5 \times 10^6 = 5 \text{ MHz.}$$

6.26 The electric field radiated by a short dipole antenna is given in spherical coordinates by

$$\begin{aligned} \mathbf{E}(R, \theta; t) &= \\ \hat{\theta} \frac{2 \times 10^{-2}}{R} \sin \theta \cos(6\pi \times 10^8 t - 2\pi R) & \quad (\text{V/m}). \end{aligned}$$

Find $\mathbf{H}(R, \theta; t)$.

Solution: Converting to phasor form, the electric field is given by

$$\tilde{\mathbf{E}}(R, \theta) = \hat{\theta} E_\theta = \hat{\theta} \frac{2 \times 10^{-2}}{R} \sin \theta \exp -j2\pi R \quad (\text{V/m}),$$

which can be used with Eq. (6.87) to find the magnetic field:

$$\begin{aligned} \tilde{\mathbf{H}}(R, \theta) &= \frac{1}{-j\omega\mu} \nabla \times \tilde{\mathbf{E}} = \frac{1}{-j\omega\mu} \left[\hat{\mathbf{R}} \frac{1}{R \sin \theta} \frac{\partial E_\theta}{\partial \phi} + \hat{\phi} \frac{1}{R} \frac{\partial}{\partial R} (R E_\theta) \right] \\ &= \frac{1}{-j\omega\mu} \hat{\phi} \frac{2 \times 10^{-2}}{R} \sin \theta \frac{\partial}{\partial R} (\exp -j2\pi R) \\ &= \hat{\phi} \frac{2\pi}{6\pi \times 10^8 \times 4\pi \times 10^{-7}} \frac{2 \times 10^{-2}}{R} \sin \theta \exp -j2\pi R \\ &= \hat{\phi} \frac{53}{R} \sin \theta \exp -j2\pi R \quad (\mu\text{A/m}). \end{aligned}$$

Converting back to instantaneous value, this is

$$\vec{H}(R, \theta; t) = \hat{\phi} \frac{53}{R} \sin \theta \cos(6\pi \times 10^8 t - 2\pi R) \quad (\mu\text{A/m}).$$

6.28 In free space, the magnetic field is given by

$$\mathbf{H} = \hat{\phi} \frac{36}{r} \cos(6 \times 10^9 t - kz) \quad (\text{mA/m}).$$

- (a) Determine k .
- (b) Determine \mathbf{E} .
- (c) Determine \mathbf{J}_d .

Solution:

(a) From the given expression, $\omega = 6 \times 10^9$ (rad/s), and since the medium is free space,

$$k = \frac{\omega}{c} = \frac{6 \times 10^9}{3 \times 10^8} = 20 \quad (\text{rad/m}).$$

(b) Convert \mathbf{H} to phasor:

$$\tilde{\mathbf{H}} = \hat{\phi} \frac{36}{r} e^{-jkz} \quad (\text{mA/m})$$

$$\begin{aligned} \tilde{\mathbf{E}} &= \frac{1}{j\omega\epsilon_0} \nabla \times \tilde{\mathbf{H}} \\ &= \frac{1}{j\omega\epsilon_0} \left[-\hat{\mathbf{r}} \frac{\partial H_\phi}{\partial z} + \hat{\mathbf{z}} \frac{1}{r} \frac{\partial}{\partial r} (rH_\phi) \right] \\ &= \frac{1}{j\omega\epsilon_0} \left[-\hat{\mathbf{r}} \frac{\partial}{\partial z} \left(\frac{36}{r} e^{-jkz} \right) + \hat{\mathbf{z}} \frac{\partial}{\partial r} (36e^{-jkz}) \right] \\ &= \frac{1}{j\omega\epsilon_0} \left[\hat{\mathbf{r}} \frac{j36k}{r} e^{-jkz} \right] \\ &= \hat{\mathbf{r}} \frac{36k}{\omega\epsilon_0 r} e^{-jkz} = \hat{\mathbf{r}} \frac{36 \times 377}{r} e^{-jkz} \times 10^{-3} = \hat{\mathbf{r}} \frac{13.6}{r} e^{-j20z} \quad (\text{V/m}). \end{aligned}$$

$$\begin{aligned} \mathbf{E} &= \Re\{[\tilde{\mathbf{E}}e^{j\omega t}]\} \\ &= \hat{\mathbf{r}} \frac{13.6}{r} \cos(6 \times 10^9 t - 20z) \quad (\text{V/m}). \end{aligned}$$

(c)

$$\begin{aligned} \mathbf{J}_d &= \epsilon_0 \frac{\partial \mathbf{E}}{\partial t} \\ &= \hat{\mathbf{r}} \frac{13.6}{r} \epsilon_0 \frac{\partial}{\partial t} (\cos(6 \times 10^9 t - 20z)) \\ &= -\hat{\mathbf{r}} \frac{13.6\epsilon_0 \times 6 \times 10^9}{r} \sin(6 \times 10^9 t - 20z) \quad (\text{A/m}^2) \\ &= -\hat{\mathbf{r}} \frac{0.72}{r} \sin(6 \times 10^9 t - 20z) \quad (\text{A/m}^2). \end{aligned}$$